

7.7 Transmission lines and waveguides

Transmission line relations

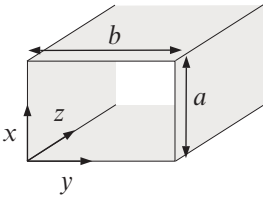
Loss-free transmission line equations	$\frac{\partial V}{\partial x} = -L \frac{\partial I}{\partial t} \quad (7.170)$ $\frac{\partial I}{\partial x} = -C \frac{\partial V}{\partial t} \quad (7.171)$	V potential difference across line I current in line L inductance per unit length C capacitance per unit length
Wave equation for a lossless transmission line	$\frac{1}{LC} \frac{\partial^2 V}{\partial x^2} = \frac{\partial^2 V}{\partial t^2} \quad (7.172)$ $\frac{1}{LC} \frac{\partial^2 I}{\partial x^2} = \frac{\partial^2 I}{\partial t^2} \quad (7.173)$	x distance along line t time
Characteristic impedance of lossless line	$Z_c = \sqrt{\frac{L}{C}} \quad (7.174)$	Z_c characteristic impedance
Characteristic impedance of lossy line	$Z_c = \sqrt{\frac{R + j\omega L}{G + j\omega C}} \quad (7.175)$	R resistance per unit length of conductor G conductance per unit length of insulator ω angular frequency
Wave speed along a lossless line	$v_p = v_g = \frac{1}{\sqrt{LC}} \quad (7.176)$	v_p phase speed v_g group speed
Input impedance of a terminated lossless line	$Z_{in} = Z_c \frac{Z_t \cos kl - jZ_c \sin kl}{Z_c \cos kl - jZ_t \sin kl} \quad (7.177)$ $= Z_c^2 / Z_t \quad \text{if } l = \lambda/4 \quad (7.178)$	Z_{in} (complex) input impedance Z_t (complex) terminating impedance k wavenumber ($= 2\pi/\lambda$)
Reflection coefficient from a terminated line	$r = \frac{Z_t - Z_c}{Z_t + Z_c} \quad (7.179)$	l distance from termination r (complex) voltage reflection coefficient
Line voltage standing wave ratio	$VSWR = \frac{1 + r }{1 - r } \quad (7.180)$	

Transmission line impedances^a

Coaxial line	$Z_c = \sqrt{\frac{\mu}{4\pi^2\epsilon}} \ln \frac{b}{a} \simeq \frac{60}{\sqrt{\epsilon_r}} \ln \frac{b}{a} \quad (7.181)$	Z_c characteristic impedance (Ω) a radius of inner conductor b radius of outer conductor ϵ permittivity ($= \epsilon_0 \epsilon_r$)
Open wire feeder	$Z_c = \sqrt{\frac{\mu}{\pi^2\epsilon}} \ln \frac{l}{r} \simeq \frac{120}{\sqrt{\epsilon_r}} \ln \frac{l}{r} \quad (7.182)$	μ permeability ($= \mu_0 \mu_r$) r radius of wires l distance between wires ($\gg r$)
Paired strip	$Z_c = \sqrt{\frac{\mu}{\epsilon}} \frac{d}{w} \simeq \frac{377}{\sqrt{\epsilon_r}} \frac{d}{w} \quad (7.183)$	d strip separation w strip width ($\gg d$)
Microstrip line	$Z_c \simeq \frac{377}{\sqrt{\epsilon_r} [(w/h) + 2]} \quad (7.184)$	h height above earth plane ($\ll w$)

^aFor lossless lines.

Waveguides^a

Waveguide equation	$k_g^2 = \frac{\omega^2}{c^2} - \frac{m^2\pi^2}{a^2} - \frac{n^2\pi^2}{b^2} \quad (7.185)$	k_g wavenumber in guide ω angular frequency a guide height b guide width m, n mode indices with respect to a and b (integers) c speed of light
Guide cutoff frequency	$v_c = c\sqrt{\left(\frac{m}{2a}\right)^2 + \left(\frac{n}{2b}\right)^2} \quad (7.186)$	v_c cutoff frequency $\omega_c = 2\pi v_c$
Phase velocity above cutoff	$v_p = \frac{c}{\sqrt{1 - (v_c/v)^2}} \quad (7.187)$	v_p phase velocity v frequency
Group velocity above cutoff	$v_g = c^2/v_p = c\sqrt{1 - (v_c/v)^2} \quad (7.188)$	v_g group velocity
Wave impedances ^b	$Z_{TM} = Z_0\sqrt{1 - (v_c/v)^2} \quad (7.189)$ $Z_{TE} = Z_0/\sqrt{1 - (v_c/v)^2} \quad (7.190)$	Z_{TM} wave impedance for transverse magnetic modes Z_{TE} wave impedance for transverse electric modes Z_0 impedance of free space ($= \sqrt{\mu_0/\epsilon_0}$)
Field solutions for TE_{mn} modes^c $B_x = \frac{\mathbf{i}k_g c^2}{\omega_c^2} \frac{\partial B_z}{\partial x} \quad E_x = \frac{\mathbf{i}\omega c^2}{\omega_c^2} \frac{\partial B_z}{\partial y} \quad (7.191)$ $B_y = \frac{\mathbf{i}k_g c^2}{\omega_c^2} \frac{\partial B_z}{\partial y} \quad E_y = \frac{-\mathbf{i}\omega c^2}{\omega_c^2} \frac{\partial B_z}{\partial x}$ $B_z = B_0 \cos \frac{m\pi x}{a} \cos \frac{n\pi y}{b} \quad E_z = 0$ Field solutions for TM_{mn} modes^c $E_x = \frac{\mathbf{i}k_g c^2}{\omega_c^2} \frac{\partial E_z}{\partial x} \quad B_x = \frac{-\mathbf{i}\omega}{\omega_c^2} \frac{\partial E_z}{\partial y} \quad (7.192)$ $E_y = \frac{\mathbf{i}k_g c^2}{\omega_c^2} \frac{\partial E_z}{\partial y} \quad B_y = \frac{\mathbf{i}\omega}{\omega_c^2} \frac{\partial E_z}{\partial x}$ $E_z = E_0 \sin \frac{m\pi x}{a} \sin \frac{n\pi y}{b} \quad B_z = 0$ 		

^aEquations are for lossless waveguides with rectangular cross sections and no dielectric.

^bThe ratio of the electric field to the magnetic field strength in the xy plane.

^cBoth TE and TM modes propagate in the z direction with a further factor of $\exp[\mathbf{i}(k_g z - \omega t)]$ on all components. B_0 and E_0 are the amplitudes of the z components of magnetic flux density and electric field respectively.